

Fundamentals of Computer Systems

Boolean Logic

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Boolean Logic

AN INVESTIGATION
OF
THE LAWS OF THOUGHT,
ON WHICH ARE FOUNDED
THE MATHEMATICAL THEORIES OF LOGIC
AND PROBABILITIES.

BY
GEORGE BOOLE, LL.D.

PROFESSOR OF MATHEMATICS IN QUEEN'S COLLEGE, COKE.

LONDON:
WALTON AND MABERLY,
UPPER GOWER-STREET, AND IVY-LANE, PATERNOSTER-ROW.
CAMBRIDGE: MACMILLAN AND CO.

1854.



George Boole
1815–1864

Boole's Intuition Behind Boolean Logic

Variables X, Y, \dots represent classes of things

No imprecision: A thing either is or is not in a class

If X is "sheep"
and Y is "white
things," XY are
all white sheep,

$$XY = YX$$

and

$$XX = X.$$

If X is "men"
and Y is
"women," $X + Y$
is "both men
and women,"

$$X + Y = Y + X$$

and

$$X + X = X.$$

If X is "men," Y
is "women," and
 Z is "European,"
 $Z(X + Y)$ is
"European men
and women"
and

$$Z(X+Y) = ZX+ZY.$$

The Axioms of (Any) Boolean Algebra

A Boolean Algebra consists of

A set of values A

An “and” operator “.”

An “or” operator “+”

A “not” operator \bar{X}

A “false” value $0 \in A$

A “true” value $1 \in A$

The Axioms of (Any) Boolean Algebra

A Boolean Algebra consists of

A set of values A

An “and” operator “ \cdot ”

An “or” operator “ $+$ ”

A “not” operator $\bar{}$

A “false” value $0 \in A$

A “true” value $1 \in A$

Axioms

$$X + Y = Y + X$$

$$X \cdot Y = Y \cdot X$$

$$X + (Y + Z) = (X + Y) + Z$$

$$X \cdot (Y \cdot Z) = (X \cdot Y) \cdot Z$$

$$X + (X \cdot Y) = X$$

$$X \cdot (X + Y) = X$$

$$X \cdot (Y + Z) = (X \cdot Y) + (X \cdot Z)$$

$$X + (Y \cdot Z) = (X + Y) \cdot (X + Z)$$

$$X + \bar{X} = 1$$

$$X \cdot \bar{X} = 0$$

The Axioms of (Any) Boolean Algebra

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A set of values A

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$$X + \bar{X} = 1$$

$$X \cdot \bar{X} = 0$$

We will use the first non-trivial Boolean Algebra:

$A = \{0, 1\}$. This adds the law of excluded middle: if $X \neq 0$ then $X = 1$ and if $X \neq 1$ then $X = 0$.

Simplifying a Boolean Expression

“You are a New Yorker if you were born in New York or were not born in New York and lived here ten years.”

$$X + (\bar{X} \cdot Y)$$

Axioms

$$X + Y = Y + X$$

$$X \cdot Y = Y \cdot X$$

$$X + (Y + Z) = (X + Y) + Z$$

$$X \cdot (Y \cdot Z) = (X \cdot Y) \cdot Z$$

$$X + (X \cdot Y) = X$$

$$X \cdot (X + Y) = X$$

$$X \cdot (Y + Z) = (X \cdot Y) + (X \cdot Z)$$

$$X + (Y \cdot Z) = (X + Y) \cdot (X + Z)$$

$$X + \bar{X} = 1$$

$$X \cdot \bar{X} = 0$$

Lemma:

$$\begin{aligned} X \cdot 1 &= X \cdot (X + \bar{X}) \\ &= X \cdot (X + Y) \text{ if } Y = \bar{X} \\ &= X \end{aligned}$$

Simplifying a Boolean Expression

“You are a New Yorker if you were born in New York or were not born in New York and lived here ten years.”

$$\begin{aligned} X + (\bar{X} \cdot Y) \\ = (X + \bar{X}) \cdot (X + Y) \end{aligned}$$

Axioms

$$\begin{aligned} X + Y &= Y + X \\ X \cdot Y &= Y \cdot X \\ X + (Y + Z) &= (X + Y) + Z \\ X \cdot (Y \cdot Z) &= (X \cdot Y) \cdot Z \\ X + (X \cdot Y) &= X \\ X \cdot (X + Y) &= X \\ X \cdot (Y + Z) &= (X \cdot Y) + (X \cdot Z) \\ X + (Y \cdot Z) &= (X + Y) \cdot (X + Z) \\ X + \bar{X} &= 1 \\ X \cdot \bar{X} &= 0 \end{aligned}$$

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Axioms

$$\begin{aligned} X + Y &= Y + X \\ X \cdot Y &= Y \cdot X \\ X + (Y + Z) &= (X + Y) + Z \\ X \cdot (Y \cdot Z) &= (X \cdot Y) \cdot Z \\ X + (X \cdot Y) &= X \\ X \cdot (X + Y) &= X \\ X \cdot (Y + Z) &= (X \cdot Y) + (X \cdot Z) \\ X + (Y \cdot Z) &= (X + Y) \cdot (X + Z) \\ X + \bar{X} &= 1 \\ X \cdot \bar{X} &= 0 \end{aligned}$$

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Simplifying a Boolean Expression

“You are a New Yorker if you were born in New York or were not born in New York and lived here ten years.”

$$\begin{aligned} X + (\bar{X} \cdot Y) \\ &= (X + \bar{X}) \cdot (X + Y) \\ &= 1 \cdot (X + Y) \\ &= X + Y \end{aligned}$$

Axioms

$$\begin{aligned} X + Y &= Y + X \\ X \cdot Y &= Y \cdot X \\ X + (Y + Z) &= (X + Y) + Z \\ X \cdot (Y \cdot Z) &= (X \cdot Y) \cdot Z \\ X + (X \cdot Y) &= X \\ X \cdot (X + Y) &= X \\ X \cdot (Y + Z) &= (X \cdot Y) + (X \cdot Z) \\ X + (Y \cdot Z) &= (X + Y) \cdot (X + Z) \\ X + \bar{X} &= 1 \\ X \cdot \bar{X} &= 0 \end{aligned}$$

Lemma:

$$\begin{aligned} X \cdot 1 &= X \cdot (X + \bar{X}) \\ &= X \cdot (X + Y) \text{ if } Y = \bar{X} \\ &= X \end{aligned}$$

More properties

$$0 + 0 = 0$$

$$0 + 1 = 1$$

$$1 + 0 = 1$$

$$1 + 1 = 1$$

$$1 + 1 + \dots + 1 = 1$$

$$X + 0 = X$$

$$X + 1 = 1$$

$$X + X = X$$

$$X + XY = X$$

$$X + \overline{X}Y = X + Y$$

$$0 \cdot 0 = 0$$

$$0 \cdot 1 = 0$$

$$1 \cdot 0 = 0$$

$$1 \cdot 1 = 1$$

$$1 \cdot 1 \cdot \dots \cdot 1 = 1$$

$$X \cdot 0 = 0$$

$$X \cdot 1 = X$$

$$X \cdot X = X$$

$$X \cdot (X + Y) = X$$

$$X \cdot (\overline{X} + Y) = XY$$

More Examples (1)

$$\begin{aligned}XY + YZ(Y + Z) &= XY + YZY + YZZ \\ &= XY + YZ \\ &= Y(X + Z)\end{aligned}$$

$$\begin{aligned}X + Y(X + Z) + XZ &= X + YX + YZ + XZ \\ &= X + YZ + XZ \\ &= X + YZ\end{aligned}$$

More Examples (2)

$$\begin{aligned}XYZ + X(\bar{Y} + \bar{Z}) &= XYZ + X\bar{Y} + X\bar{Z} && \text{Expand} \\ &= X(YZ + \bar{Y} + \bar{Z}) && \text{Factor w.r.t. } X \\ &= X(YZ + \bar{Y} + \bar{Z} + Y\bar{Z}) && \bar{Z} \rightarrow \bar{Z} + Y\bar{Z} \\ &= X(YZ + Y\bar{Z} + \bar{Y} + \bar{Z}) && \text{Reorder} \\ &= X(Y(Z + \bar{Z}) + \bar{Y} + \bar{Z}) && \text{Factor w.r.t. } Y \\ &= X(Y + \bar{Y} + \bar{Z}) && Y + \bar{Y} = 1 \\ &= X(1 + \bar{Z}) && 1 + \bar{Z} = 1 \\ &= X && X1 = X\end{aligned}$$

$$\begin{aligned}(X + \bar{Y} + \bar{Z})(X + \bar{Y}Z) &= XX + X\bar{Y}Z + \bar{Y}X + \bar{Y}\bar{Y}Z + \bar{Z}X + \bar{Z}\bar{Y}Z \\ &= X + X\bar{Y}Z + X\bar{Y} + \bar{Y}Z + X\bar{Z} \\ &= X + \bar{Y}Z\end{aligned}$$

Sum-of-products form

Can always reduce a complex Boolean expression to a sum of product terms:

$$\begin{aligned}XY + \bar{X}(X + Y(Z + X\bar{Y}) + \bar{Z}) &= XY + \bar{X}(X + YZ + YX\bar{Y} + \bar{Z}) \\ &= XY + \bar{X}X + \bar{X}YZ + \bar{X}YX\bar{Y} + \bar{X}\bar{Z} \\ &= XY + \bar{X}YZ + \bar{X}\bar{Z} \\ &\quad \text{(can do better)} \\ &= Y(X + \bar{X}Z) + \bar{X}\bar{Z} \\ &= Y(X + Z) + \bar{X}\bar{Z} \\ &= \overline{Y\bar{X}\bar{Z}} + \bar{X}\bar{Z} \\ &= Y + \bar{X}\bar{Z}\end{aligned}$$

What Does This Have To Do With Logic Circuits?

A SYMBOLIC ANALYSIS
OF
RELAY AND SWITCHING CIRCUITS

by

Claude Elwood Shannon
B.S., University of Michigan
1936

Submitted in Partial Fulfillment of the
Requirements for the Degree of
MASTER OF SCIENCE
from the
Massachusetts Institute of Technology
1940

Signature of Author _____
Department of Electrical Engineering, August 10, 1937

Signature of Professor _____
in Charge of Research

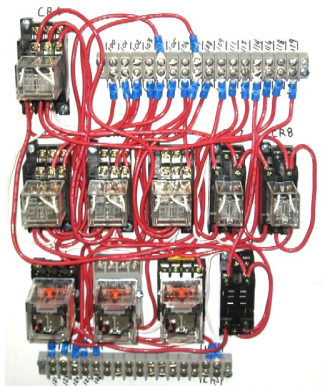
Signature of Chairman of Department _____
Committee on Graduate Students



Claude Shannon
1916–2001

Shannon's MS Thesis

“We shall limit our treatment to circuits containing only relay contacts and switches, and therefore at any given time the circuit between any two terminals must be either open (infinite impedance) or closed (zero impedance).”



Shannon's MS Thesis

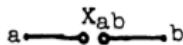


Fig. 1

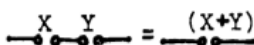


Fig. 2

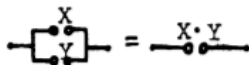


Fig. 3

"It is evident that with the above definitions the following postulates hold.

$$0 \cdot 0 = 0$$

A closed circuit in parallel with a closed circuit is a closed circuit.

$$1 + 1 = 1$$

An open circuit in series with an open circuit is an open circuit.

$$1 + 0 = 0 + 1 = 1$$

An open circuit in series with a closed circuit in either order is an open circuit.

$$0 \cdot 1 = 1 \cdot 0 = 0$$

A closed circuit in parallel with an open circuit in either order is an open circuit.

$$0 + 0 = 0$$





A closed circuit in series with a closed circuit is a closed circuit.

$$1 \cdot 1 = 1$$

An open circuit in parallel with an open circuit is an open circuit.

At any give time either $X = 0$ or $X = 1$

Alternate Notations for Boolean Logic

Operator	Math	Engineer	Schematic
Copy	x	X	x — or x —  — x
Complement	$\neg x$	\bar{X}	x —  — \bar{x}
AND	$x \wedge y$	XY or $X \cdot Y$	X —  — XY Y —
OR	$x \vee y$	$X + Y$	X —  — $X + Y$ Y —

Definitions

Literal: a Boolean variable or its complement

E.g., X \bar{X} Y \bar{Y}

Implicant: A product of literals

E.g., X XY $X\bar{Y}Z$

Minterm: An implicant with each variable once

E.g., $X\bar{Y}Z$ XYZ $\bar{X}\bar{Y}Z$

Maxterm: A sum of literals with each variable once

E.g., $X + \bar{Y} + Z$ $X + Y + Z$ $\bar{X} + \bar{Y} + Z$

Be Careful with Bars

$$\overline{XY} \neq \overline{XY}$$

Be Careful with Bars

$$\overline{X\overline{Y}} \neq \overline{X\overline{Y}}$$

Let's check all the combinations of X and Y :

X	Y	\overline{X}	\overline{Y}	$\overline{X \cdot Y}$	XY	$\overline{X\overline{Y}}$
0	0	1	1	1	0	1
0	1	1	0	0	0	1
1	0	0	1	0	0	1
1	1	0	0	0	1	0

Truth Tables

A *truth table* is a canonical representation of a Boolean function

X	Y	XY
0	0	0
0	1	0
1	0	0
1	1	1

Minterms

Each row has a unique minterm

X	Y	Minterm	$\overline{X}\overline{Y}$	$\overline{X}Y$	$X\overline{Y}$	XY
0	0	$\overline{X}\overline{Y}$	1	0	0	0
0	1	$\overline{X}Y$	0	1	0	0
1	0	$X\overline{Y}$	0	0	1	0
1	1	XY	0	0	0	1

The minterm is the product term that is 1 for only its row

Maxterms

Each row has a unique maxterm

X	Y	Maxterm	$X+Y$	$X+\bar{Y}$	$\bar{X}+Y$	$\bar{X}+\bar{Y}$
0	0	$X+Y$	0	1	1	1
0	1	$X+\bar{Y}$	1	0	1	1
1	0	$\bar{X}+Y$	1	1	0	1
1	1	$\bar{X}+\bar{Y}$	1	1	1	0

The maxterm is the sum term that is 1 for only its row

Sum-of-minterms and Product-of-maxterms

Two mechanical ways to translate a function's truth table into an expression:

X	Y	Minterm	Maxterm	F
0	0	$\bar{X}\bar{Y}$	$X+Y$	0
0	1	$\bar{X}Y$	$X+\bar{Y}$	1
1	0	$X\bar{Y}$	$\bar{X}+Y$	1
1	1	XY	$\bar{X}+\bar{Y}$	0

The sum of the minterms where the function is 1:

$$F = \bar{X}Y + X\bar{Y}$$

The product of the maxterms where the function is 0:

$$F = (X+Y)(\bar{X}+\bar{Y})$$

Expressions to Schematics

$$F = \bar{X}Y + X\bar{Y}$$

X

Y

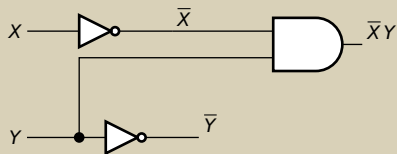
Expressions to Schematics

$$F = \bar{X}Y + X\bar{Y}$$



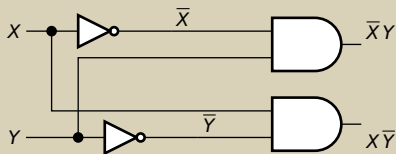
Expressions to Schematics

$$F = \bar{X}Y + X\bar{Y}$$



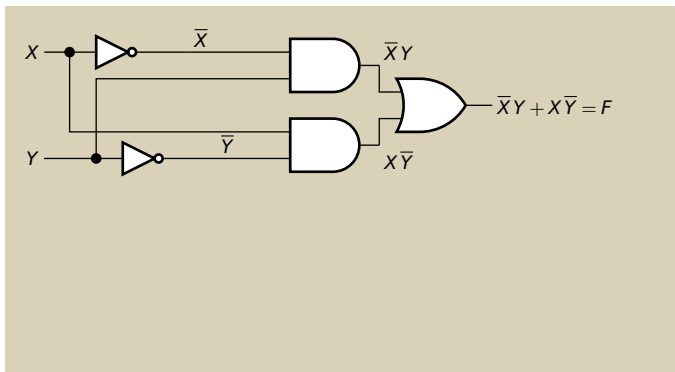
Expressions to Schematics

$$F = \bar{X}Y + X\bar{Y}$$



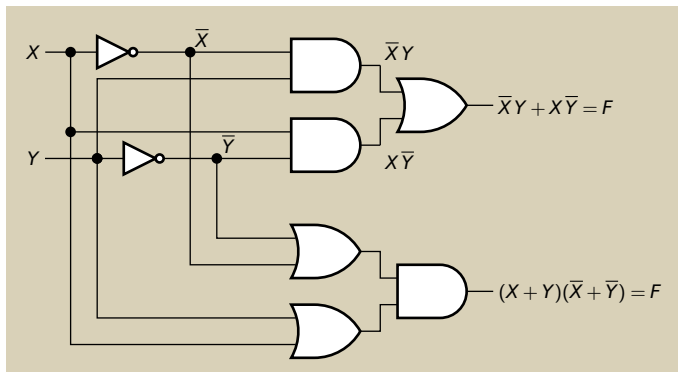
Expressions to Schematics

$$F = \bar{X}Y + X\bar{Y}$$



Expressions to Schematics

$$F = \bar{X}Y + X\bar{Y} = (X + Y)(\bar{X} + \bar{Y})$$



Minterms and Maxterms: Another Example

The minterm and maxterm representation of functions may look very different:

X	Y	Minterm	Maxterm	F
0	0	$\overline{X}\overline{Y}$	$X+Y$	0
0	1	$\overline{X}Y$	$X+\overline{Y}$	1
1	0	$X\overline{Y}$	$\overline{X}+Y$	1
1	1	XY	$\overline{X}+\overline{Y}$	1

The sum of the minterms where the function is 1:

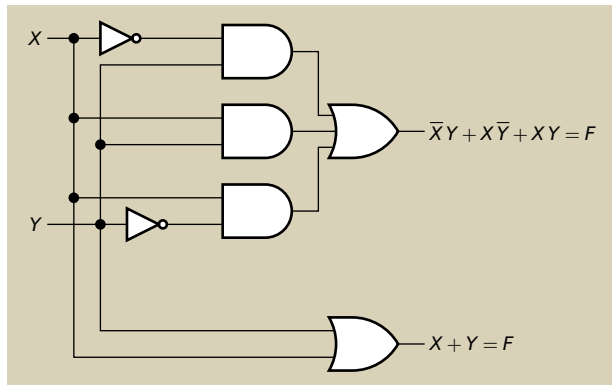
$$F = \overline{X}Y + X\overline{Y} + XY$$

The product of the maxterms where the function is 0:

$$F = X + Y$$

Expressions to Schematics 2

$$F = \bar{X}Y + X\bar{Y} + XY = X + Y$$



The Menagerie of Gates



The Menagerie of Gates

Buffer



0		0
1		1

Inverter



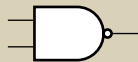
0		1
1		0

AND



·		0	1
0		0	0
1		0	1

NAND



·		0	1
0		1	1
1		1	0

OR



+		0	1
0		0	1
1		1	1

NOR



$\bar{+}$		0	1
0		1	0
1		0	0

XOR



\oplus		0	1
0		0	1
1		1	0

XNOR



$\bar{\oplus}$		0	1
0		1	0
1		0	1

De Morgan's Theorem

$$\overline{X+Y} = \bar{X} \cdot \bar{Y}$$

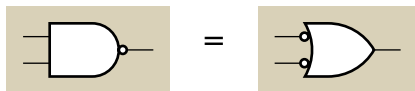
$$\overline{X \cdot Y} = \bar{X} + \bar{Y}$$

Proof by Truth Table:

X	Y	$X+Y$	$\bar{X} \cdot \bar{Y}$	$X \cdot Y$	$\bar{X} + \bar{Y}$
0	0	0	1	0	1
0	1	1	0	0	1
1	0	1	0	0	1
1	1	1	0	1	0

De Morgan's Theorem in Gates

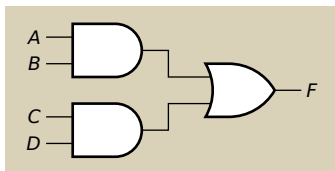
$$\overline{AB} = \overline{A} + \overline{B}$$



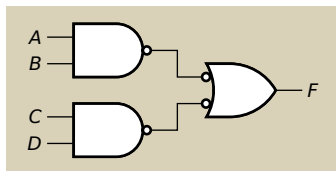
$$\overline{A + B} = \overline{A} \cdot \overline{B}$$



Bubble Pushing

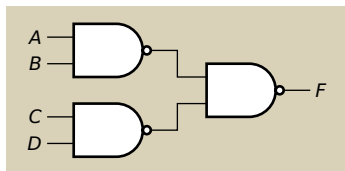


Bubble Pushing



Two bubbles on a wire cancel

Bubble Pushing

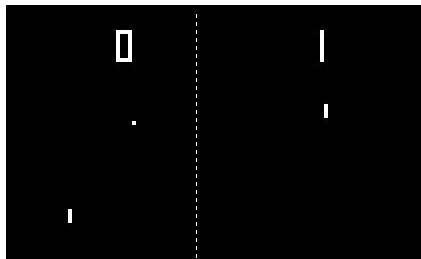


Two bubbles on a wire cancel

Apply De Morgan's Theorem (i.e., push the bubbles through the gates)

Transform OR with inverted inputs into NAND

PONG



PONG, Atari 1973

Built from TTL logic gates; no computer, no software

Launched the video arcade game revolution

Horizontal Ball Control in PONG

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>
0	0	0	X	X
0	0	1	0	1
0	1	0	0	1
0	1	1	X	X
1	0	0	X	X
1	0	1	1	0
1	1	0	1	1
1	1	1	X	X

The ball moves either left or right.

Part of the control circuit has three inputs: *M* (“move”), *L* (“left”), and *R* (“right”).

It produces two outputs *A* and *B*.

Here, “X” means “I don’t care what the output is; I never expect this input combination to occur.”

Horizontal Ball Control in PONG

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>
0	0	0	0	0
0	0	1	0	1
0	1	0	0	1
0	1	1	0	0
1	0	0	0	0
1	0	1	1	0
1	1	0	1	1
1	1	1	0	0

E.g., assume all the X's are 0's and use Minterms:

$$A = M\bar{L}R + ML\bar{R}$$

$$B = \bar{M}\bar{L}R + \bar{M}L\bar{R} + ML\bar{R}$$

3 inv + 4 AND3 + 1 OR2 + 1 OR3

Horizontal Ball Control in PONG

M	L	R	A	B
0	0	0	1	1
0	0	1	0	1
0	1	0	0	1
0	1	1	1	1
1	0	0	1	1
1	0	1	1	0
1	1	0	1	1
1	1	1	1	1

Assume all the X's are 1's and use Maxterms:

$$A = (M + L + \bar{R})(M + \bar{L} + R)$$

$$B = \bar{M} + L + \bar{R}$$

3 inv + 3 OR3 + 1 AND2

Horizontal Ball Control in PONG

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>
0	0	0	0	1
0	0	1	0	1
0	1	0	0	1
0	1	1	0	1
1	0	0	1	1
1	0	1	1	0
1	1	0	1	1
1	1	1	1	0

Choosing better values for the X's
and being much more clever:

$$A = M$$

$$B = \overline{MR}$$

1 NAND2 (!)

Karnaugh Maps

Basic trick: put “similar” variable values near each other so simple functions are obvious

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>
0	0	0	X	X
0	0	1	0	1
0	1	0	0	1
0	1	1	X	X
1	0	0	X	X
1	0	1	1	0
1	1	0	1	1
1	1	1	X	X

The *M*'s are already arranged nicely

Karnaugh Maps

Basic trick: put “similar” variable values near each other so simple functions are obvious

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>
0	0	0	X	X
0	0	1	0	1
0	1	0	0	1
0	1	1	X	X
1	0	0	X	X
1	0	1	1	0
			1	1
			1	1

Let's rearrange the *L*'s by permuting two pairs of rows

1 1
X X

Karnaugh Maps

Basic trick: put “similar” variable values near each other so simple functions are obvious

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>
0	0	0	X	X
0	0	1	0	1
0	1	0	0	1
0	1	1	X	X
1	0	0	X	X
1	0	1	1	0

Let's rearrange the *L*'s by permuting two pairs of rows

1	1	0	1	1
1	1	1	X	X

Karnaugh Maps

Basic trick: put “similar” variable values near each other so simple functions are obvious

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>
0	0	0	X	X
0	0	1	0	1
0	1	0	0	1
0	1	1	X	X
1	0	0	X	X
1	0	1	1	0

Let's rearrange the *L*'s by permuting two pairs of rows

1	1	0	1	1
1	1	1	X	X

Karnaugh Maps

Basic trick: put “similar” variable values near each other so simple functions are obvious

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>
0	0	0	X	X
0	0	1	0	1
0	1	0	0	1
0	1	1	X	X
1	0	0	X	X
1	0	1	1	0

Let's rearrange the *L*'s by permuting two pairs of rows

1	1	0	1	1
1	1	1	X	X

Karnaugh Maps

Basic trick: put “similar” variable values near each other so simple functions are obvious

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>
0	0	0	X	X
0	0	1	0	1
0	1	0	0	1
0	1	1	X	X
1	0	0	X	X
1	0	1	1	0

Let's rearrange the *L*'s by permuting two pairs of rows

1	1	0	1	1
1	1	1	X	X

Karnaugh Maps

Basic trick: put “similar” variable values near each other so simple functions are obvious

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>			
0	0	0	X	X			
0	0	1	0	1			
0	1	0	0	1			
0	1	1	X	X			
			1	1	0	1	1
			1	1	1	X	X
1	0	0	X	X			
1	0	1	1	0			

Let's rearrange the *L*'s by permuting two pairs of rows

Karnaugh Maps

Basic trick: put “similar” variable values near each other so simple functions are obvious

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>		
0	0	0	X	X		
0	0	1	0	1		
0	1	0	0	1		
0	1	1	X	X		
		1	1	0	1	1
		1	1	1	X	X
1	0	0	X	X		
1	0	1	1	0		

Let's rearrange the *L*'s by permuting two pairs of rows

Karnaugh Maps

Basic trick: put “similar” variable values near each other so simple functions are obvious

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>
0	0	0	X	X
0	0	1	0	1
0	1	0	0	1
0	1	1	X	X
1	1	0	1	1
1	1	1	X	X
1	0	0	X	X
1	0	1	1	0

Let's rearrange the *L*'s by permuting two pairs of rows

Karnaugh Maps

Basic trick: put “similar” variable values near each other so simple functions are obvious

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>
0	0	0	X	X
0	0	1	0	1
0	1	0	0	1
0	1	1	X	X
1	1	0	1	1
1	1	1	X	X
1	0	0	X	X
1	0	1	1	0

The *R*'s are really crazy; let's use the second dimension

Karnaugh Maps

Basic trick: put “similar” variable values near each other so simple functions are obvious

M	L	R	A	B
0 ₀	0 ₀	0 ₁	X ₀	X ₁
0 ₀	1 ₁	0 ₁	0 _X	1 _X
1 ₁	1 ₁	0 ₁	1 _X	1 _X
1 ₁	0 ₀	0 ₁	X ₁	X ₀

The R 's are really crazy; let's use the second dimension

Karnaugh Maps

Basic trick: put “similar” variable values near each other so simple functions are obvious

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>
00	00	01	X0	X1
00	11	01	0X	1X
11	11	01	1X	1X
11	00	01	X1	X0

The *R*'s are really crazy; let's use the second dimension

Karnaugh Maps

Basic trick: put “similar” variable values near each other so simple functions are obvious

<i>M</i>	<i>L</i>	<i>R</i>	<i>A</i>	<i>B</i>
00	00	01	X0	X1
00	11	01	0X	1X
11	11	01	1X	1X
11	00	01	X1	X0

MR

M

Maurice Karnaugh's Maps

The Map Method for Synthesis of Combinational Logic Circuits

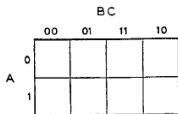
M. KARNAUGH

NONMEMBER AIEE

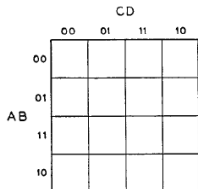
THE SEARCH for simple abstract techniques to be applied to the design of switching systems is still, despite some recent advances, in its early stages. The problem in this area which has been attacked most energetically is that of the synthesis of efficient combinational that is, nonsequential, logic circuits.

be convenient to describe other methods in terms of Boolean algebra. Whenever the term "algebra" is used in this paper, it will refer to Boolean algebra, where addition corresponds to the logical connective "or," while multiplication corresponds to "and."

The minimizing chart,³ developed at



(A)



(B)

Fig. 2. Graphical representations of the input conditions for three and for four variables

Karnaugh maps (a.k.a., K-maps)

All functions can be expressed with a map. There is one square for each minterm in a function's truth table

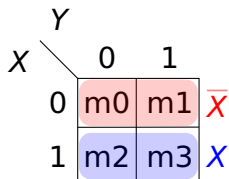
X	Y	minterm	
0	0	$\bar{X}\bar{Y}$	m0
0	1	$\bar{X}Y$	m1
1	0	$X\bar{Y}$	m2
1	1	XY	m3

		Y	
		0	1
X	0	m0	m1
	1	m2	m3

Karnaugh maps (a.k.a., K-maps)

All functions can be expressed with a map. There is one square for each minterm in a function's truth table

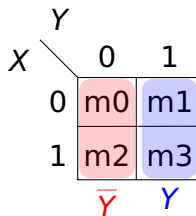
X	Y	minterm	
0	0	$\bar{X}\bar{Y}$	m0
0	1	$\bar{X}Y$	m1
1	0	$X\bar{Y}$	m2
1	1	XY	m3



Karnaugh maps (a.k.a., K-maps)

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X	Y	minterm	
0	0	$\bar{X}\bar{Y}$	m0
0	1	$\bar{X}Y$	m1
1	0	$X\bar{Y}$	m2
1	1	XY	m3



Karnaugh maps (a.k.a., K-maps) – Cont. 1

Fill out the table with the values of some function.

X	Y	F
0	0	0
0	1	1
1	0	1
1	1	1

		Y	
		0	1
X	0	0	1
	1	1	1

Karnaugh maps (a.k.a., K-maps) – Cont. 2

When two cells share an edge and both are 1, those two terms can be combined to form a single, simpler term.

		Y	
		0	1
X	0	0	1
	1	1	1

$$F = \overline{X}Y + X\overline{Y} + XY$$

Karnaugh maps (a.k.a., K-maps) – Cont. 2

When two cells share an edge and both are 1, those two terms can be combined to form a single, simpler term.

	Y	0	1
X	0	0	1
1	1	1	1

$$F = \bar{X}Y + X\bar{Y} + XY$$

	Y	0	1
X	0	0	1
1	1	1	1

$$F = X + \bar{X}Y$$

	Y	0	1
X	0	0	1
1	1	1	1

$$F = Y + X\bar{Y}$$

Karnaugh maps (a.k.a., K-maps) – Cont. 2

When two cells share an edge and both are 1, those two terms can be combined to form a single, simpler term.

	Y	0	1
X	0	0	1
1	1	1	1

$$F = \bar{X}Y + X\bar{Y} + XY$$

	Y	0	1
X	0	0	1
1	1	1	1

$$F = X + \bar{X}Y$$

	Y	0	1
X	0	0	1
1	1	1	1

$$F = Y + X\bar{Y}$$

	Y	0	1
X	0	0	1
1	1	1	1

$$F = X + Y$$

Karnaugh maps (a.k.a., K-maps) – Summary So Far

- ▶ Circle contiguous groups of 1s (circle sizes must be a power of 2)
- ▶ There is a correspondence between circles on a k-map and terms in a function expression
- ▶ The bigger the circle, the simpler the term
- ▶ Add circles (and terms) until all 1s on the k-map are circled
- ▶ Prime implicant: circles that can be no bigger (smallest product term)
- ▶ Essential prime implicant: circles that uniquely covers a 1 is “essential”

		Y	
		0	1
X	0	0	1
	1	1	1

$$F = X + Y$$

3-Variable Karnaugh Maps

- ▶ Use gray ordering on edges with multiple variables
- ▶ Gray encoding: order of values such that only one bit changes at a time
- ▶ Two minterms are considered adjacent if they differ in only one variable (this means maps “wrap”)

		Y Z			
		00	01	11	10
X	0	m0	m1	m3	m2
	1	m4	m5	m7	m6

3-Variable Karnaugh Maps

- ▶ Use gray ordering on edges with multiple variables
- ▶ Gray encoding: order of values such that only one bit changes at a time
- ▶ Two minterms are considered adjacent if they differ in only one variable (this means maps “wrap”)

		Y Z			
		00	01	11	10
X	0	m0	m1	m3	m2
	1	m4	m5	m7	m6

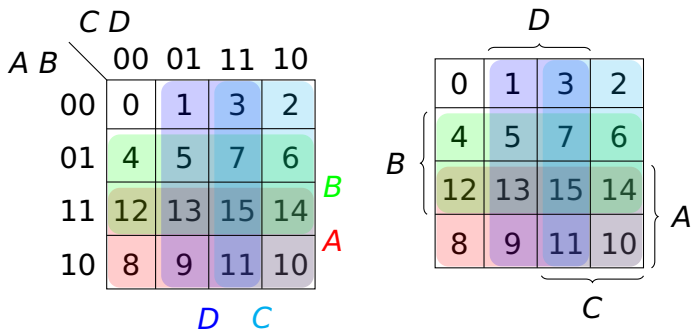
Z Y

X

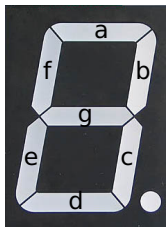
		Z			
		m0	m1	m3	m2
X	{	m4	m5	m7	m6
			Y		

4-Variable Karnaugh Maps

An extension of 3-variable maps.



The Seven-Segment Decoder Example



<i>W</i>	<i>X</i>	<i>Y</i>	<i>Z</i>	<i>a</i>	<i>b</i>	<i>c</i>	<i>d</i>	<i>e</i>	<i>f</i>	<i>g</i>
0	0	0	0	1	1	1	1	1	1	0
0	0	0	1	0	1	1	0	0	0	0
0	0	1	0	1	1	0	1	1	0	1
0	0	1	1	1	1	1	1	0	0	1
0	1	0	0	0	1	1	0	0	1	1
0	1	0	1	1	0	1	1	0	1	1
0	1	1	0	1	0	1	1	1	1	1
0	1	1	1	1	1	1	0	0	0	0
1	0	0	0	1	1	1	1	1	1	1
1	0	0	1	1	1	1	0	0	1	1
1	0	1	0	X	X	X	X	X	X	X
1	0	1	1	X	X	X	X	X	X	X
1	1	0	0	X	X	X	X	X	X	X
1	1	0	1	X	X	X	X	X	X	X
1	1	1	0	X	X	X	X	X	X	X
1	1	1	1	0	0	0	0	0	0	0

Karnaugh Map for Seg. a

W	X	Y	Z	a
0	0	0	0	1
0	0	0	1	0
0	0	1	0	1
0	0	1	1	1
0	1	0	0	0
0	1	0	1	1
0	1	1	0	1
0	1	1	1	1
1	0	0	0	1
1	0	0	1	1
1	0	1	0	X
1	0	1	1	X
1	1	0	0	X
1	1	0	1	X
1	1	1	0	X
1	1	1	1	0

		Z				
		1	0	1	1	
X	{	0	1	1	1	}
		X	X	0	X	
		1	1	X	X	
				Y		

The Karnaugh Map Sum-of-Products Challenge

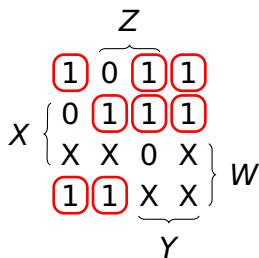
Cover all the 1's and none of the 0's using **as few literals** (gate inputs) as possible.

Few, large rectangles are good.

Covering X's is optional.

Karnaugh Map for Seg. a

W	X	Y	Z	a
0	0	0	0	1
0	0	0	1	0
0	0	1	0	1
0	0	1	1	1
0	1	0	0	0
0	1	0	1	1
0	1	1	0	1
0	1	1	1	1
1	0	0	0	1
1	0	0	1	1
1	0	1	0	X
1	0	1	1	X
1	1	0	0	X
1	1	0	1	X
1	1	1	0	X
1	1	1	1	0



The minterm solution: cover each 1 with a single implicant.

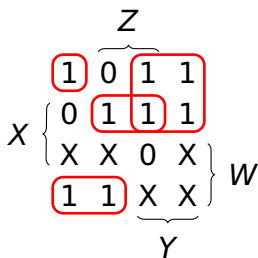
$$\begin{aligned}
 a = & \overline{W}\overline{X}\overline{Y}\overline{Z} + \overline{W}\overline{X}YZ + \overline{W}\overline{X}Y\overline{Z} + \\
 & \overline{W}X\overline{Y}Z + \overline{W}XYZ + \overline{W}XY\overline{Z} + \\
 & W\overline{X}\overline{Y}\overline{Z} + W\overline{X}\overline{Y}Z
 \end{aligned}$$

$8 \times 4 = 32$ literals

4 inv + 8 AND4 + 1 OR8

Karnaugh Map for Seg. a

W	X	Y	Z	a
0	0	0	0	1
0	0	0	1	0
0	0	1	0	1
0	0	1	1	1
0	1	0	0	0
0	1	0	1	1
0	1	1	0	1
0	1	1	1	1
1	0	0	0	1
1	0	0	1	1
1	0	1	0	X
1	0	1	1	X
1	1	0	0	X
1	1	0	1	X
1	1	1	0	X
1	1	1	1	0



Merging implicants helps

Recall the distributive law:

$$AB + AC = A(B + C)$$

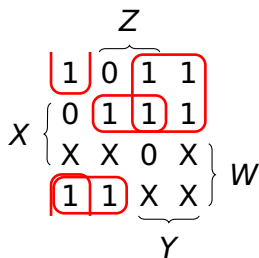
$$a = \overline{W}\overline{X}\overline{Y}\overline{Z} + \overline{W}Y + \overline{W}XZ + W\overline{X}\overline{Y}$$

$$4 + 2 + 3 + 3 = 12 \text{ literals}$$

$$4 \text{ inv} + 1 \text{ AND}_4 + 2 \text{ AND}_3 + 1 \text{ AND}_2 + 1 \text{ OR}_4$$

Karnaugh Map for Seg. a

W	X	Y	Z	a
0	0	0	0	1
0	0	0	1	0
0	0	1	0	1
0	0	1	1	1
0	1	0	0	0
0	1	0	1	1
0	1	1	0	1
0	1	1	1	1
1	0	0	0	1
1	0	0	1	1
1	0	1	0	X
1	0	1	1	X
1	1	0	0	X
1	1	0	1	X
1	1	1	0	X
1	1	1	1	0



Missed one: Remember this is actually a torus.

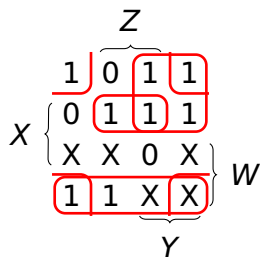
$$a = \overline{X}\overline{Y}\overline{Z} + \overline{W}Y + \overline{W}XZ + W\overline{X}\overline{Y}$$

3 + 2 + 3 + 3 = 11 literals

4 inv + 3 AND3 + 1 AND2 + 1 OR4

Karnaugh Map for Seg. a

W	X	Y	Z	a
0	0	0	0	1
0	0	0	1	0
0	0	1	0	1
0	0	1	1	1
0	1	0	0	0
0	1	0	1	1
0	1	1	0	1
0	1	1	1	1
1	0	0	0	1
1	0	0	1	1
1	0	1	0	X
1	0	1	1	X
1	1	0	0	X
1	1	0	1	X
1	1	1	0	X
1	1	1	1	0



Taking don't-cares into account, we can enlarge two implicants:

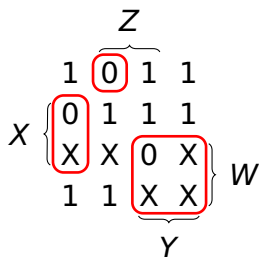
$$a = \overline{X}\overline{Z} + \overline{W}Y + \overline{W}XZ + W\overline{X}$$

2 + 2 + 3 + 2 = 9 literals

3 inv + 1 AND3 + 3 AND2 + 1 OR4

Karnaugh Map for Seg. a

W	X	Y	Z	a
0	0	0	0	1
0	0	0	1	0
0	0	1	0	1
0	0	1	1	1
0	1	0	0	0
0	1	0	1	1
0	1	1	0	1
0	1	1	1	1
1	0	0	0	1
1	0	0	1	1
1	0	1	0	X
1	0	1	1	X
1	1	0	0	X
1	1	0	1	X
1	1	1	0	X
1	1	1	1	0



Can also compute the complement of the function and invert the result.

Covering the 0's instead of the 1's:

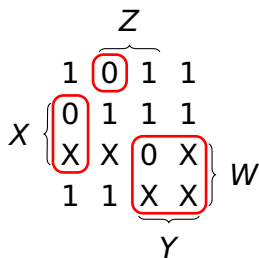
$$\bar{a} = \bar{W}\bar{X}\bar{Y}Z + X\bar{Y}\bar{Z} + WY$$

4 + 3 + 2 = 9 literals

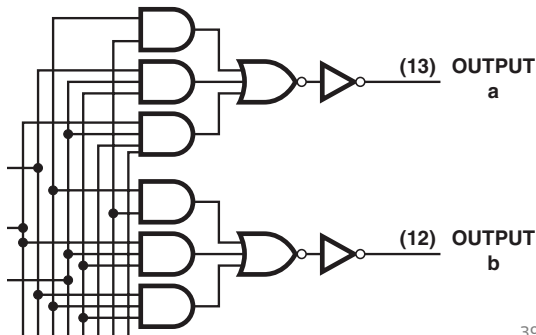
5 inv + 1 AND4 + 1 AND3 + 1 AND2
+ 1 OR3

Karnaugh Map for Seg. a

W	X	Y	Z	a
0	0	0	0	1
0	0	0	1	0
0	0	1	0	1
0	0	1	1	1
0	1	0	0	0
0	1	0	1	1
0	1	1	0	1
0	1	1	1	1
1	0	0	0	1
1	0	0	1	1
1	0	1	0	X
1	0	1	1	X
1	1	0	0	X
1	1	0	1	X
1	1	1	0	X
1	1	1	1	0



To display the score, PONG used a TTL chip with this solution in it:



Boolean Laws and Karnaugh Maps

		W			
Y	{	0	0	1	1
		0	0	1	1
		0	0	1	1
		0	0	1	1
				Z	
		X			

$$\begin{aligned} &WXY\bar{Z} + \bar{W}XY\bar{Z} + \\ &WXYZ + \bar{W}XYZ + \\ &WX\bar{Y}\bar{Z} + \bar{W}X\bar{Y}\bar{Z} + \\ &WX\bar{Y}Z + \bar{W}X\bar{Y}Z \end{aligned}$$

Factor out the W 's

Boolean Laws and Karnaugh Maps

		W			
		0	0	1	1
Y	{	0	0	1	1
		0	0	1	1
		0	0	1	1
		0	0	1	1
		X			

$$\begin{aligned} &(W + \bar{W})X\bar{Y}\bar{Z} + \\ &(W + \bar{W})XY\bar{Z} + \\ &(W + \bar{W})XYZ + \\ &(W + \bar{W})X\bar{Y}Z \end{aligned}$$

Use the identities

$$W + \bar{W} = 1$$

and

$$1X = X.$$

Boolean Laws and Karnaugh Maps

	W					
	0	0	1	1		
Y	{	0	0	1	1	}
		0	0	1	1	
		0	0	1	1	
		0	0	1	1	
			X			

$X\bar{Y}\bar{Z}+$
 $XY\bar{Z}+$
 $XYZ+$
 $X\bar{Y}Z$

Factor out the Y's

Boolean Laws and Karnaugh Maps

	W						
	0	0	1	1			
Y	{	0	0	1	1	}	Z
	0	0	1	1			
	0	0	1	1			
	0	0	1	1			
			X				

$$(\bar{Y} + Y)X\bar{Z} +$$

$$(\bar{Y} + Y)XZ$$

Apply the identities again

Boolean Laws and Karnaugh Maps

	W					
	0	0	1	1		
Y	{	0	0	1	1	}
		0	0	1	1	
		0	0	1	1	
		0	0	1	1	
			X			

$$X\bar{Z} +$$
$$XZ$$

Factor out Z

Boolean Laws and Karnaugh Maps

	W				
	0	0	1	1	
Y	0	0	1	1	
	0	0	1	1	
	0	0	1	1	Z
	0	0	1	1	
			X		

$$X(\bar{Z} + Z)$$

Simplify

Boolean Laws and Karnaugh Maps

	W					
	0	0	1	1		
Y	{	0	0	1	1	}
		0	0	1	1	
		0	0	1	1	
		0	0	1	1	
			X			

X
Done